

Set 1– due 26 January

“The world of our imagination is narrower and more special in its logical structure than the world of physical things.” (Max Born)

1) [10 points] Jackson 9.3. In CGS, the discontinuity in $E \cdot \hat{n}$ across a conductor is $4\pi\sigma$ where σ is the surface charge density. Or, just do the problems in MKS and let us deal with the units...

2) [20 points] Consider an array of $2N + 1$ dipoles. Each has a time-dependent dipole moment $\vec{p}(t) = \hat{z}pe^{-i\omega t}$, and they are spaced a distance D apart along the \hat{z} axis, centered on the origin. What is the resulting antenna pattern? Explicitly work out what happens at θ (the polar angle measured with respect to the \hat{z} axis) near $\pi/2$.

This is an iconic problem: many similar radiators distributed in space, so the antenna pattern tells you things about the spatial distribution of radiators.

3) (a) [15 points] The surface of a charge distribution of uniform density is almost a sphere: the radius as a function of the polar angle is $R(\theta) = R_0(1 + \gamma \cos \theta)$. The quantity γ oscillates in time with frequency ω , so we are describing something like surface waves on a water balloon. Working to lowest nontrivial order in the small parameter γ , and in the long wavelength limit, find an expression for the antenna pattern of emitted radiation and the total power radiated.
(b) [5 points] Part (a) is a variation of Jackson 9.12, except in that problem $R(\theta) = R_0(1 + \beta P_2(\cos \theta))$. What is the leading multipole behavior of the radiation in that case? Why?

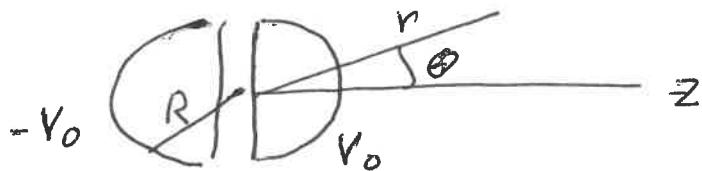
Jackson 9.3. The potential between the hemispheres

9.3.1

is

$$V(t) = V_0 \cos \omega t = \operatorname{Re} V_0 e^{-i\omega t}$$

Since these are conductors, a charge is induced on them, and there is an induced time dependent dipole moment which radiates.



$$\begin{aligned} V(r, \theta) &= \sum_l a_l \left(\frac{r}{R}\right)^l P_l(\cos \theta) \text{ if } r < R \\ &= \sum_l b_l \left(\frac{R}{r}\right)^{l+1} P_l(\cos \theta) \text{ if } r > R \end{aligned}$$

V is continuous at $r=R$ so $a_l = b_l$. In CGS

$$4\pi \sigma = - \left. \frac{\partial V}{\partial r} \right|_{r=R^+} + \left. \frac{\partial V}{\partial r} \right|_{r=R^-} \quad (1)$$

We only want the dipole term:

$$4\pi \sigma_D = \frac{a_1 + 2b_1}{R} \cos \theta = \frac{3a_1}{R} \cos \theta \quad (2)$$

$$\int V(R, \theta) P_l(\cos \theta) d\cos \theta = \frac{2}{2l+1} a_l$$

$$V_0 \left[\int_0^1 \cos \theta d\cos \theta - \int_{-1}^0 \cos \theta d\cos \theta \right] = \frac{2}{3} a_1$$

$$\text{or } V_0 = \frac{2}{3} a_1 \text{ or } a_1 = \frac{3}{2} V_0 \text{ so}$$

$$\sigma_D = \frac{3}{4\pi R} \cdot \frac{3}{2} V_0 \cos \theta$$

$$\text{The dipole moment } \vec{d} = \hat{z} \int d\cdot R \cos\theta dA \quad 9.3.2$$

$$\begin{aligned}\vec{P} &= \hat{z} \frac{q}{8\pi} V_0 \frac{R^3}{R} \int_{-1}^1 d\cos\theta \cos^2\theta \cdot 2\pi \\ &= \hat{z} V_0 R^2 \cdot \frac{q}{8\pi} \cdot 2\pi \cdot \frac{2}{3} = \frac{3}{2} V_0 R^2 \hat{z} \end{aligned} \quad (3)$$

and we are done! $\vec{B} = k^2 (\hat{n} \times \vec{p}) \frac{e^{ikr}}{r}$
 $\vec{E} = -\hat{n} \times \vec{B}$

power $P = \frac{c k^4}{3} |\vec{p}|^2 = \frac{3}{4} c V_0^2 (k R)^4$

In MKS $\sigma_D / \epsilon_0 = \frac{3a_1}{R} \cos\theta \quad (2')$

$a_1 = \frac{3}{2} V_0$, the same as for CGS;

$$\sigma_D = \frac{q}{2} \frac{V_0}{R} \epsilon_0 \cos\theta, \quad \vec{p} = \hat{z} \cdot \frac{3}{2} V_0 R^2 \epsilon_0 \cdot 2\pi \cdot \frac{2}{3}$$

$$\vec{p} = \frac{3}{2} V_0 R^2 (4\pi \epsilon_0) \hat{z} \Rightarrow 6\pi \epsilon_0 V_0 R^2 \hat{z}$$

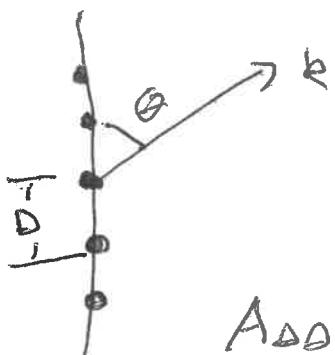
Power radiated = $\frac{c^2 Z_0 k^4}{12\pi} (6\pi \epsilon_0 V_0 R^2)^2$

$$P = 3\pi V_0^2 (k R)^4 \epsilon^2 Z_0 \epsilon_0^2.$$

$$\text{With } Z_0 = \sqrt{\frac{\mu_0}{\epsilon_0}} \rightarrow c^2 Z_0 \epsilon_0^2 = \frac{\epsilon_0^2}{\mu_0 \epsilon_0} \sqrt{\frac{\mu_0}{\epsilon_0}} = \sqrt{\frac{\epsilon_0}{\mu_0}} = \frac{1}{Z_0}$$

and $P = \frac{V_0^2}{Z_0} \cdot 3\pi (k R)^4 \cdot Z_0$ or resistance,
so the units are correct.

2) An array of dipoles, the d th source offset by $\vec{r}_d = \hat{z}(dD)$ for $d = -N$ to N .



The vector potential is

$$\vec{A} = \frac{e^{ikr}}{r} \int d^3x J(x) e^{ik \cdot x}$$

Assume that J peaks near $\vec{x} = \vec{x}_d$. Write $\vec{x} = \vec{x}_d + \vec{y}$. Assume the range of y is much less than x_d (or Δx_d). Then

$$A_d \sim \int d^3x J(x) e^{ik \cdot x} = e^{ik \cdot x_d} \int d^3y J(y) e^{ik \cdot y}$$

and $\vec{A}_d = \left[\frac{e^{ikr}}{r} \right] \vec{p} e^{ik \cdot \vec{x}_d}$.

Here \vec{p} is the individual dipole moment measured with respect to location \vec{x}_d , or to $\vec{y} = 0$.

For all the dipoles

$$\vec{A} = \left[\frac{e^{ikr}}{r} \right] \hat{z} p \sum_{d=-N}^N e^{ik \cdot \vec{r}_d} = \frac{e^{ikr}}{r} \hat{z} p F(\hat{n}).$$

The antenna pattern is

$$\frac{dP}{d\Omega} = \frac{dP_0}{d\Omega} |F(\hat{n})|^2 \text{ where } \frac{dP_0}{d\Omega} \text{ is the}$$

antenna pattern for a single dipole, $\frac{dP_0}{d\Omega} = \frac{8\pi}{3} k^4 p^2 \sin^2$

Note the factorization - this occurs again & again dealing with the physics of similar objects.

We need to compute the form factor:

$$F(\vec{h}) = \sum_{j=-N}^N \exp i j D k \vec{h} \cdot \hat{\vec{z}} = \sum_{j=-N}^N \exp i j D k \cos \theta.$$

Defining $x = \exp ikD \cos \theta$,

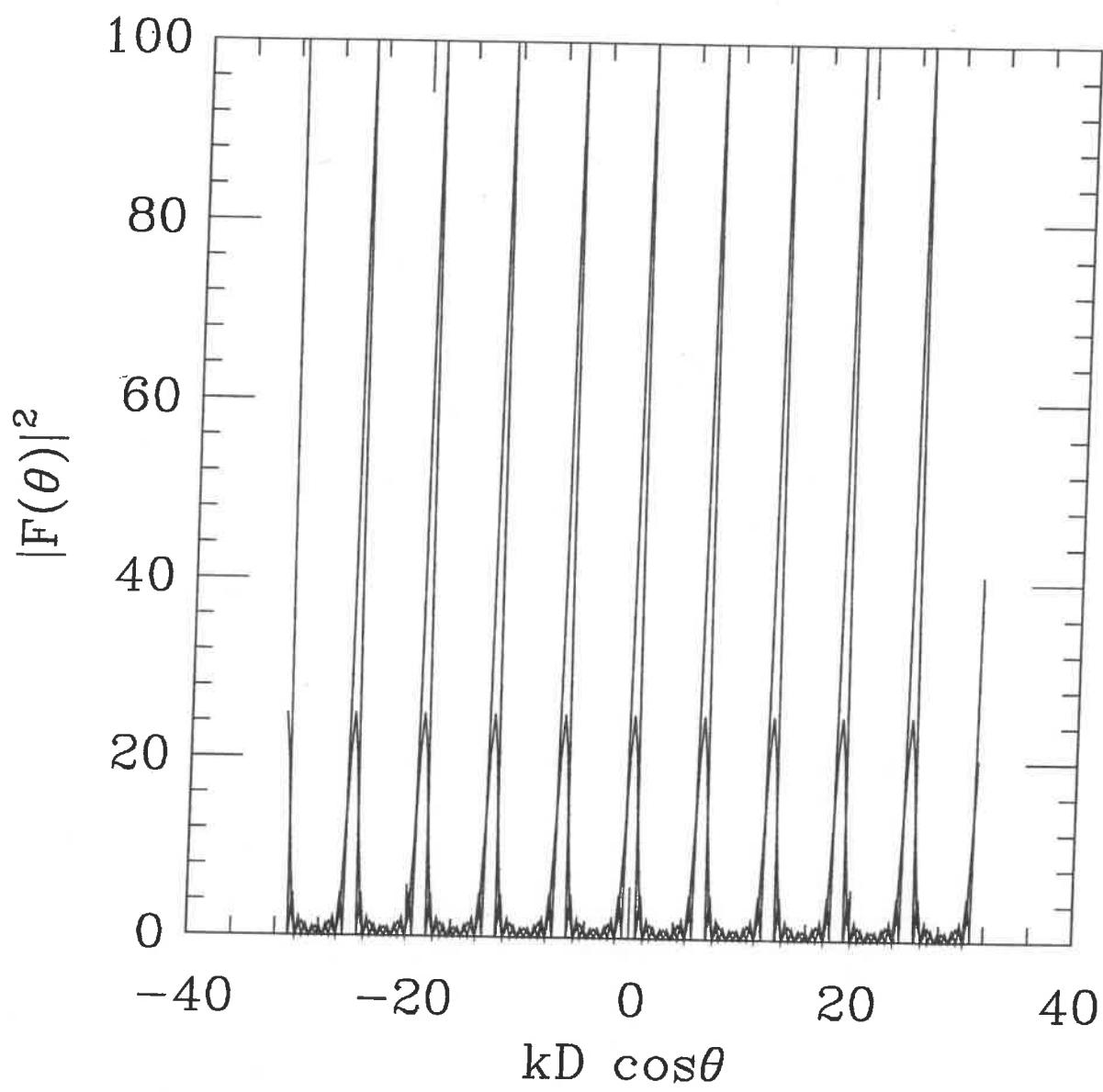
$$\begin{aligned} F(\vec{h}) &= \sum_{j=-N}^N x^j = x^{-N} \sum_{k=0}^{2N} x^k \\ &= x^{-N} \left(\sum_{k=0}^{\infty} x^k - x^{2N+1} \sum_{k=0}^{\infty} x^k \right) \\ &= x^{-N} \left[\frac{1 - x^{2N+1}}{1 - x} \right] = \left(\frac{x^{-N} - x^{N+1}}{1 - x} \right) \int \frac{dx}{x} \\ &= \frac{x^{-N-\frac{1}{2}} - x^{N+\frac{1}{2}}}{\frac{1}{\sqrt{x}} - \sqrt{x}} = \frac{e^{-i(N+\frac{1}{2})k \cos \theta} - e^{i(N+\frac{1}{2})k \cos \theta}}{e^{-i \frac{kD}{2} \cos \theta} - e^{i \frac{kD}{2} \cos \theta}} \\ &= \frac{\sin \left[(N+\frac{1}{2}) k D \cos \theta \right]}{\sin \left[\frac{k D}{2} \cos \theta \right]} \end{aligned}$$

(see plot on next page!)

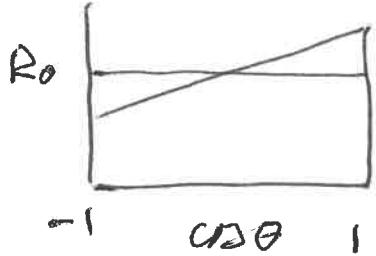
Near $\theta = \pi/2$ this is roughly $\frac{N+\frac{1}{2}}{\frac{1}{2}} = 2N+1$

and $\frac{dP}{d\Omega} = (2N+1)^2 \frac{df_0}{d\Omega}$ - this is coherent radiation from the $2N+1$ antennas.

$$2N+1 = 5 \times 10$$



$$3) R(\theta) = R_0 [1 + 8 \cos \theta] \text{ and } 8 = 8_0 e^{-i\omega t} \quad R-1$$



The volume of the charge is

$$\int dV = \int_0^{2\pi} d\phi \int_{-1}^1 d\cos\theta \int_0^{R(\theta)} r^2 dr.$$

For small δ the total charge is

$$\begin{aligned} Q &= \epsilon_0 \int dV = \epsilon_0 \cdot \frac{2\pi}{3} R_0^3 \int_{-1}^1 d\cos\theta [1 + 8 \cos\theta]^3 \\ &= \frac{2\pi}{3} R_0^3 \epsilon_0 \left[2 + 3 \int_{-1}^1 d\cos\theta (8 \cos\theta + 8^2 \cos^2\theta) \right. \\ &\quad \left. + \dots \right] \\ &= \frac{4\pi R_0^3}{3} \epsilon_0 + O(\delta^2) \end{aligned}$$

The lowest multipole is the electric dipole. By symmetry, $P_x = P_y = 0$ while

$$\begin{aligned} P_z &= 2\pi \epsilon_0 \int_{-1}^1 d\cos\theta \int_0^{R(\theta)} r^2 dr \cdot r \cos\theta \\ &= \frac{2\pi}{4} \epsilon_0 R_0^4 \int_{-1}^1 d\cos\theta [1 + 8 \cos\theta]^4 \cos^2\theta \end{aligned}$$

To leading order in δ ,

$$\begin{aligned} P_z &= \frac{2\pi}{4} R_0^4 \left[\frac{3Q}{4\pi R_0^3} \right] \cdot 48 \int_{-1}^1 \cos^2\theta d\cos\theta \\ &= 2 \cdot \frac{3}{4} Q R_0 \delta \cdot \frac{2}{3} = Q R_0 \delta. \end{aligned}$$

This goes into the dipole formula

In CGS

$$\frac{dP}{d\Omega} = \frac{c k^4 |P_z|^2}{8\pi} \sin^2 \theta$$

$$P = \frac{c}{3} k^4 |P_z|^2 = \frac{c k^4}{3} [Q R_0 \delta]^2$$

In MKS

$$\frac{dP}{d\Omega} = \frac{c^2 Z_0}{32\pi^2} k^4 |P_z|^2 \sin^2 \theta$$

$$P = \frac{c^2 Z_0}{12\pi} k^4 |P_z|^2$$

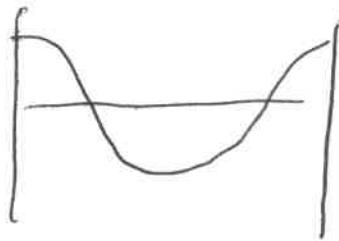
either way, $\frac{1}{P} \frac{dP}{d\Omega} = \frac{3}{8\pi} \sin^2 \theta.$

b) In Jackson 9.12, $R(\theta) = 1 + \beta P_2(\cos \theta)$

$$\text{and } P_2(\cos \theta) = \frac{3 \cos^2 \theta - 1}{2}.$$

The shape remains symmetric about the equator.

This will give rise to quadrupole radiation.



$\approx QDD$